

Minimum Pearson Distance Detection for Multi-Level Channels with Gain and/or Offset Mismatch

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Channels with gain and offset mismatch

Let $\mathbf{x} = (x_1, \dots, x_n)$ be a codeword of n symbols over the alphabet $\mathcal{Q} = \{0, 1, \dots, q - 1\}$ taken from a codebook S .

In AWGN channels, it is assumed that if \mathbf{x} is sent that $\mathbf{r} = \mathbf{x} + \boldsymbol{\nu}$ is received, where $\boldsymbol{\nu}$ is an additive Gaussian, $N(0, \sigma)$, noise term vector.

Here we assume an AWGN offset and gain *mismatch* channel, i.e.,

$$\mathbf{r} = a(\mathbf{x} + \boldsymbol{\nu}) + b,$$

where $a > 0$ and b are constants, $a > 0$.

Detection is normally very sensitive to gain and offset mismatch, and a loss of performance may occur.

Prior Art solutions

- Automatic offset and gain control based on the previous words received. (Not possible for Flash memories.)
- Insertion of reference symbols by which actual gain and offset can be estimated.
- Intrinsically resistant codes.

Apply Minimum Euclidean Distance Detection

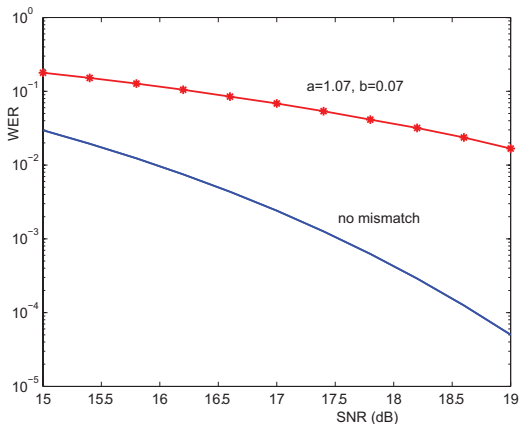
$$x_o = \arg \min_{\hat{x} \in S} d(\hat{x}),$$

where

$$d(\hat{x}) = \sum_{i=1}^n (r_i - \hat{x}_i)^2 = \sum_{i=1}^n (ax_i + b - \hat{x}_i)^2$$

$$d(\hat{x}) = \sum (ax_i + b)^2 - 2a \sum x_i \hat{x}_i - 2b \sum \hat{x}_i + \sum \hat{x}_i^2$$

Mismatch



Word error rate of matched and mismatched channel ($a = 1.07$, $b = 0.07$), where $n = 8$ and $q = 4$.

All codewords in a *balanced code* must satisfy

$$\sum \hat{x}_i = d_c$$

and

$$\sum \hat{x}_i^2 = d_p$$

then

$$d'(\hat{\mathbf{x}}) = - \sum r_i \hat{x}_i,$$

is independent of a and b .

Disadvantage: High redundancy of balanced codes.

New Approach: Minimum Pearson distance detection

Apply Minimum Pearson Distance Detection

$$\mathbf{x}_o = \arg \min_{\hat{\mathbf{x}} \in S} \delta(\hat{\mathbf{x}}),$$

where

$$\delta(\hat{\mathbf{x}}) = 1 - \rho_{\mathbf{r}, \hat{\mathbf{x}}}, \quad (1)$$

where

$$\rho_{\mathbf{r}, \hat{\mathbf{x}}} = \frac{\sum_{i=1}^n (r_i - \bar{r})(\hat{x}_i - \bar{\hat{x}})}{\sigma_r \sigma_{\hat{x}}} \quad (2)$$

is the well-known (*Pearson*) *correlation coefficient*. We define

$$\bar{\hat{x}} = \frac{1}{n} \sum_{i=1}^n \hat{x}_i, \quad \sigma_{\hat{x}}^2 = \sum_{i=1}^n (\hat{x}_i - \bar{\hat{x}})^2.$$

Minimum Pearson distance detection II

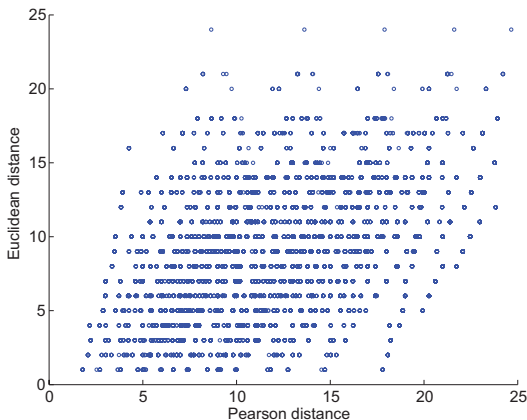
The Pearson distance, $1 - \rho_{\mathbf{r}, \hat{\mathbf{x}}}$, is invariant to changes in translation or scale (up to a sign) in the two vectors \mathbf{r} and $\hat{\mathbf{x}}$. That is,

$$\rho_{\mathbf{r}, \hat{\mathbf{x}}} = \rho_{c_1 + c_2 \mathbf{r}, \hat{\mathbf{x}}},$$

where c_1 and $c_2 \geq 0$ are real constants.

Thus: Minimum Pearson distance detection is immune to offset and gain mismatch! But some problems have to be solved.

Scatter Diagram



Scatter diagram of Euclidean distance versus Pearson distance for $n = 6$ and $q = 3$.

What is wrong and why do we need coding??

- The Pearson distance is undefined for codewords \mathbf{x} with $\sigma_{\mathbf{x}} = 0$.
- A Pearson-distance-based detector cannot distinguish between the words \mathbf{x} and $c_1 + c_2\mathbf{x}$.

Constrained coding is needed to exclude codewords

- a) $\mathbf{x} = c$ and
- b) $c_1 + c_2\mathbf{x}$ if $\mathbf{x} \in S$.

A set of T -constrained codewords is denoted by S_T consists of codewords, where T , $0 < T \leq q$, *preferred* symbols **must** appear at least once in a codeword. The size, $|S_T|$, equals

$$|S_T| = \sum_{i=0}^T (-1)^i \binom{T}{T-i} (q-i)^n, \quad n \geq T. \quad (3)$$

T-constrained codes II

Definition: The set S_1 contains codewords where the symbol '0' appears at least once.

The number of constrained codewords is

$$|S_1| = q^n - (q - 1)^n, q > 1.$$

Definition: The set S_2 contains codewords where both the symbols '0' and ' $q - 1$ ' appear at least once.

The number of constrained codewords is

$$|S_2| = q^n - 2(q - 1)^n + (q - 2)^n, q > 1.$$

For $q = 2$, we simply have

$$|S_1| = 2^n - 1 \text{ and } |S_2| = 2^n - 2.$$

The all-'0' and all-' $q - 1$ ' words are deleted.

Property A: If $\mathbf{x} \in S_1$ then $\mathbf{x} + c \notin S_1$ for all non-zero $c \in \mathbb{R}$.

Proof: By definition, \mathbf{x} has at least one position, say k , where $x_k = 0$. Hence if $c < 0$, then $x_k + c < 0$, and thus $\mathbf{x} + c \notin S_1$. If $c > 0$, then $x_i + c > 0$ for all i , and thus $\mathbf{x} + c \notin S_1$ since $\mathbf{x} + c$ does not contain the symbol '0'

Property B: If $\mathbf{x} \in S_2$ then $c_1 + c_2\mathbf{x} \notin S_2$ for all $c_1, c_2 \in \mathbb{R}$ with $(c_1, c_2) \neq (0, 1)$ and $c_2 > 0$.

Proof: Suppose $c_1 + c_2\mathbf{x} \in S_2$. Since $c_2 > 0$ and $x_i \geq 0$ for all i , it follows that $c_1 \leq 0$ in order to have at least one '0' in $c_1 + c_2\mathbf{x}$. Since $c_1 < 0$ would result in at least one negative value in $c_1 + c_2\mathbf{x}$ (in a position where \mathbf{x} has a '0'), it follows that $c_1 = 0$ and thus $c_2 \neq 1$. If $0 < c_2 < 1$, then all symbols in $c_1 + c_2\mathbf{x} = c_2\mathbf{x}$ are smaller than $q - 1$, while if $c_2 > 1$, then at least one value in $c_1 + c_2\mathbf{x}$ is larger than $q - 1$ (in a position where \mathbf{x} has a ' $q - 1$ '). In conclusion, $c_1 + c_2\mathbf{x} \notin S_2$.

Offset and gain mismatch

The word error rate (WER) is upperbounded by

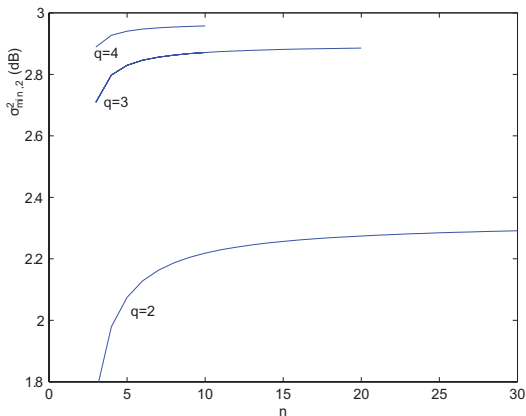
$$\text{WER} < N_1 Q\left(\frac{d_{\min,2}}{2\sigma}\right), \quad \sigma \ll 1,$$

where

$$d_{\min,2}^2 = \min_{\substack{\mathbf{x}, \hat{\mathbf{x}} \in S_2 \\ \mathbf{x} \neq \hat{\mathbf{x}}}} 2\sigma_x^2(1 - \rho_{\mathbf{x}, \hat{\mathbf{x}}}),$$

and N_1 is a constant.

Minimum squared distance, $d_{\min,2}^2$



Detection loss, $-20 \log(d_{\min,2})$ (dB), as a function of the codeword length n with $q = 2, 3, 4$ as a parameter. Although lines have been drawn, the curves consist of discrete points.

Here we assume an *AWGN offset mismatch* channel, i.e.,

$$\mathbf{r} = \mathbf{x} + \boldsymbol{\nu} + b,$$

where b a constant. We propose the distance measure

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) = \sum_{i=1}^n (r_i - \hat{x}_i + \bar{\hat{x}})^2, \quad (4)$$

where, as we may notice, we have deleted $\sigma_{\hat{x}}$, in (1). We first show that $\delta_1(\mathbf{r}, \hat{\mathbf{x}})$ is, as claimed, independent of the channel's offset b

Analysis of offset mismatch case

$$\begin{aligned}\delta_1(\mathbf{r}, \hat{\mathbf{x}}) &= \sum_{i=1}^n (x'_i + b - \hat{x}_i + \bar{x})^2 \\ &= \sum_{i=1}^n (x'_i - \hat{x}_i + \bar{x})^2 + 2b \sum_{i=1}^n (x'_i - \hat{x}_i + \bar{x}) + b^2,\end{aligned}$$

where $x'_i = x_i + \nu_i$. By definition, we have

$$\sum_{i=1}^n (x'_i - \hat{x}_i + \bar{x}) = \sum_{i=1}^n x'_i,$$

so that

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) \equiv \sum_{i=1}^n (x'_i - \hat{x}_i + \bar{x})^2. \quad (5)$$

It is immediate that the minimization of $\delta_1(\mathbf{r}, \hat{\mathbf{x}})$ is *intrinsically resistant* to the channel offset b .

We have

$$\mathbf{x}_o = \arg \min_{\hat{\mathbf{x}} \in \mathcal{S}} \delta_1(\mathbf{r}, \hat{\mathbf{x}}). \quad (6)$$

A scaled (with a positive factor) or translated version of the metric is said to be *equivalent* with the original version of the metric.

Metric equivalence will be denoted by the \equiv sign since the outcome of (6) is unchanged. Thus,

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) \equiv c_1 \delta_1(\mathbf{r}, \hat{\mathbf{x}}) + c_2,$$

where $c_1 > 0$ and c_2 are constants.

The receiver decides that the codeword \mathbf{x}_o was sent if (4) attains its least value for $\hat{\mathbf{x}} = \mathbf{x}_o$, that is

$$\mathbf{x}_o = \arg \min_{\hat{\mathbf{x}} \in S_1} \delta_1(\mathbf{r}, \hat{\mathbf{x}}) = \arg \min_{\hat{\mathbf{x}} \in S_1} \sum_{i=1}^n (r_i - \hat{x}_i + \bar{\hat{x}})^2. \quad (7)$$

We have $r_i = x_i + \nu_i$, so that

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) = \sum_{i=1}^n (x_i + \nu_i - \hat{x}_i + \bar{\hat{x}})^2.$$

Performance Analysis II

The analysis is simplified by noticing that

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) \equiv \sum_{i=1}^n (x_i + \nu_i - \hat{x}_i + \bar{x} - \bar{x})^2.$$

Define $\mathbf{e} = \mathbf{x} - \hat{\mathbf{x}}$ and $\bar{e} = \bar{x} - \bar{\hat{x}}$, then

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) = \sum_{i=1}^n (e_i - \bar{e} + \nu_i)^2.$$

The detector errs, i.e. $\mathbf{x}_o \neq \mathbf{x}$, if there is at least one codeword $\hat{\mathbf{x}} \in S_1, \hat{\mathbf{x}} \neq \mathbf{x}$, such that

$$\delta_1(\mathbf{r}, \hat{\mathbf{x}}) < \delta_1(\mathbf{r}, \mathbf{x}),$$

or

$$\sum_{i=1}^n (e_i - \bar{e} + \nu_i)^2 < \sum_{i=1}^n \nu_i^2 \quad (8)$$

or

$$2 \sum_{i=1}^n \nu_i (e_i - \bar{e}) + \sum_{i=1}^n (e_i - \bar{e})^2 < 0. \quad (9)$$

The left-hand side of (9) is a stochastic variable with distribution $N(\alpha_1, \beta_1 \sigma^2)$, where

$$\alpha_1 = \sum_{i=1}^n (e_i - \bar{e})^2$$

and

$$\beta_1 = 4 \sum_{i=1}^n (e_i - \bar{e})^2.$$

We define the square of the distance between the vectors \mathbf{x} and $\hat{\mathbf{x}}$ by

$$d_1^2(\mathbf{x}, \hat{\mathbf{x}}) = \frac{4\alpha_1^2}{\beta_1} = \sum_{i=1}^n (e_i - \bar{e})^2.$$

The probability that $\delta_1(\mathbf{r}, \hat{\mathbf{x}}) < \delta_1(\mathbf{r}, \mathbf{x})$ equals

$$Q\left(\frac{d_1(\mathbf{x}, \hat{\mathbf{x}})}{2\sigma}\right),$$

where the Q -function is defined by

$$Q(x) = \frac{1}{\sqrt{2\pi}} \int_x^{\infty} e^{-\frac{u^2}{2}} du.$$

The word error rate (WER) is upperbounded by

$$\text{WER} < \frac{1}{|S_1|} \sum_{\mathbf{x} \in S_1} \sum_{\hat{\mathbf{x}} \neq \mathbf{x}} Q\left(\frac{d_1(\mathbf{x}, \hat{\mathbf{x}})}{2\sigma}\right). \quad (10)$$

Define the square of the minimum distance between any possible pair of codewords in S_1 by

$$d_{\min,1}^2 = \min_{\substack{\mathbf{x}, \hat{\mathbf{x}} \in S_1 \\ \mathbf{x} \neq \hat{\mathbf{x}}}} d_1^2(\mathbf{x}, \hat{\mathbf{x}}) = \min_{\mathbf{e} \neq \mathbf{0}} \sum_{i=1}^n (e_i - \bar{e})^2. \quad (11)$$

The WER is overbounded by

$$\text{WER} < N_1 Q\left(\frac{d_{\min,1}}{2\sigma}\right), \sigma \ll 1, \quad (12)$$

where N_1 is the average number of pairs of codewords (neighbors) at minimum distance, $d_{\min,1}$.

We simply find

$$d_{\min,1} = \sqrt{1 - \frac{1}{n}}, \quad (13)$$

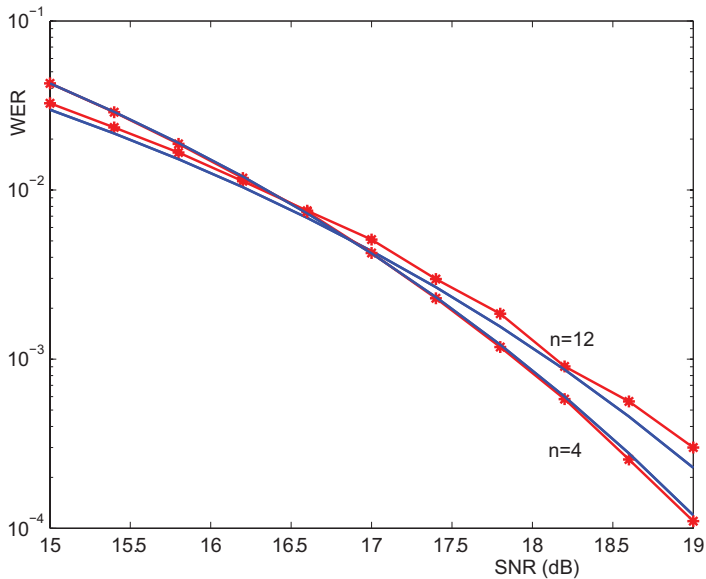
and we obtain an upperbound to the word error rate (WER)

$$\text{WER} < \frac{2n(q-1)}{q} Q\left(\frac{1}{2\sigma} \sqrt{1 - \frac{1}{n}}\right). \quad (14)$$

The figure shows the WER, computed using the above upperbound (14), as a function of the signal-to-noise-ratio (SNR), where the quantity SNR is defined by

$$\text{SNR}(\text{dB}) = -20 \log_{10} \sigma.$$

Performance of new system



- Is the T -constrained code the only class of codes that can be used with Pearson detection? Are there larger sets?

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- Can we improve the noise margin by extra coding?
- How to efficiently encode and decode T -constrained codes?
- Can we combine the Euclidean and Pearson distance metrics?

- Minimum Pearson distance detection is intrinsically resistant against gain and/or offset mismatch.
- The detection is used with T -constrained codes, where in each codeword $T = 2$ the '0' and $q - 1$ symbols appear at least once.
- The redundancy of T -constrained codes is much less than that of balanced codes of the same length.
- We have analyzed the error performance in the presence of additive Gaussian noise.